

METHOD OF CONTROLLING THE SPEED OF AN ELECTRIC MOTOR

Field of the Invention

The invention relates to a method of controlling the speed of an electric motor that is powered by a semiconductor device of the triac type, by controlling
5 the angle of the triggering signal applied to the triac.

Background of the Invention

Induction motors are single-phase motors with a series
10 excitation commutator and are frequently used in the field of domestic electrical appliances, for example for driving the rotating drum of a clothes washing or drying machine. Such motors are often powered by a semiconductor triac device. It is often desired to
15 control the speed of such a motor to increase the efficiency of its operation.

Known methods for controlling the speed of rotation of a universal motor use a tachometric generator coupled to the motor to supply an output electrical signal
20 that is representative of the motor speed and a triac, whose triggering angle determines the RMS (root mean square) voltage applied to the motor windings. Digital control methods are used to control the motor speed based on the principal that a triggering angle
25 difference of the triac is proportional to the motor speed difference.

One such control method is disclosed in U.S. Patent 6,633,149 in which a method for digital control of a universal motor, in particular for electrical
30 household appliances, measures the motor rotation speed, determines the difference between the measured speed and a preset (desired) speed, and controls the motor on the basis of this difference. This method further estimates at least one of the values of the
35 resisting torque of the motor and the current in the

motor windings. While such a method is useful, it is relatively complicated to implement particularly in household appliances, such as a laundry machine, where reduced cost is a paramount goal. Accordingly, a need
5 exists for a method of controlling the speed of a motor powered by a triac that is easy to perform and relatively inexpensive to implement.

Summary of the Invention

In accordance with the invention, a novel method is
10 provided for controlling the speed of a motor that is powered by a triac device to achieve a desired preset speed. According to the invention, a digital method is provided in which the speed of the motor is measured and a signal is produced from which a digital
15 numerical value representative of the mathematical first derivative of motor speed is derived. A determination is made of the digital numerical value of the motor speed derivative relative to a range, or band, of values of the motor speed derivative. An
20 error signal is also computed that is proportional to the error between the measured current motor speed and the preset speed. A determination is made of the digital numerical value of the motor speed error signal relative to a range, or band, of values that
25 includes a value corresponding to that of the motor operating at the preset speed.

If the value of the motor speed derivative related signal or motor speed error signal is outside of the value of its respective range or band, which
30 corresponds to an unrealistic value of motor speed, then the microprocessor program forces the value to zero. This avoids making a motor speed correction of an unrealistic value. Therefore, the motor speed error to be corrected to achieve the preset speed will
35 always be in a range of realistic values.

If the value of each of the two signals is within its respective range or band, then the two numerical values are added to produce a total error signal whose value is then converted to a signal to correct the triac triggering angle to that needed to obtain the preset motor speed. Provisions also are made to prevent over control of the triac triggering angle so that it will not become unstable.

Brief Description of the Drawings

Other objects and advantages of the present invention will become more apparent upon reference to the following description and annexed drawings, in which: Figure 1 is a schematic block diagram showing an application of the present invention; Figures 2 and 3 are flow charts showing the production of signals to be used in computing the triac triggering angle correction; and Figure 4 is a flow chart showing the production of the signal that corrects the triac triggering angle to achieve the preset motor speed.

Detailed Description of the Invention

Figure 1 shows a device 1, such as the rotating tub of a household appliance washing machine that is rotated by an electric motor 3. Motor 3 is powered by a conventional triac semiconductor device 4 that operates from a suitable power supply (not shown). In such triac devices, a triggering signal is applied to its gate electrode. The electrical angle, or time, of application of the triggering signal controls the triac conduction time and thereby its output power which in turn controls the speed of motor 3. All of this is well known in the art.

A tachometer 5 of a conventional construction is connected, such as to the rotating shaft of the motor 3, to measure the motor speed and produce a signal

that corresponds to the current motor speed. The current motor speed signal produced by the tachometer 5 preferably is in digital form and is applied to the input of a controller 7. The controller 7 is a device, such as a conventional microprocessor, that can perform calculations and has a memory section for storing data. The controller 7 is also programmed with the necessary data, such as preset motor speed and triac triggering angle needed to achieve said preset speed, as well as instructions to perform various steps, described below, and to produce an output signal. In the present invention, the controller 7 output signal is applied to a digital to analog converter 8 that produces an electrical signal applied to the triac gate electrode to control its triggering angle.

As explained above, the speed of motor 3 is to be controlled to a preset speed by setting the electrical angle of triggering the triac 4 that supplies the electrical current to operate the motor. The control of the motor speed in accordance with the invention makes use of a range, or band, of the mathematical first derivative of the motor speed. The invention also makes use of a range, or band, of a mathematical value that is proportional to the difference, or error, between the current motor speed and the preset speed. The former range is hereafter called the Derivative Band and the latter range the Proportional Band.

Figures 2 and 3 depict how the Derivative Band and Proportional Band signal values are derived by the controller 7 for use in controlling the speed of motor 3. In figure 2, the production S200 of the motor speed derivative signal value is described. The first step S201 is to calculate the value of the mathematical

first derivative of the motor current speed, which can be expressed as ds/dt . This is calculated by the controller 7 first storing the digital value of the current motor speed measured by the tachometer 5 at one time and subtracting this value from the value of measured current motor speed. This quantity is divided by the elapsed time between the two speed measurements to produce a digital quantity corresponding to the first derivative of the motor speed.

A value, hereafter called Yield D, is next computed. This is done by comparing the motor speed derivative value determined in S201 against a range of Yield D values of the motor speed derivative that is stored in the controller 7. The Yield D values are in a range from 0 to high with a median value. These are numerical digital values stored in a table in the memory section of controller 7. In S203, the motor speed derivative value determined in S201 is compared with both the current motor speed, used in S201, and the Yield D stored range of values. In the determination in S203, if both the motor speed derivative value is higher than the stored value range and the motor current speed is greater than zero (S201), then the Yield D value is output as 0 in S204. That is, in this case the motor speed derivative is in the high end of the Band D range, but not too high, and outside of the range of being a realistic value. An S204 Yield D value of 0 is output to S211. In S211, the Band D value is calculated as:

Band D = Derivative (S201) x Yield D (1)

If both the S203 conditions are met, then the final Band D value (S211) would be zero, since the S203 output value is zero.

If the conditions of S203 are not met, then the Yield D value determination passes to S205. In S205, there

is a test for a condition of the presence of a high numerical value for the motor speed derivative of S201. If this condition is met, then in S206 a digital value corresponding to Yield D= High is output to S211
5 for the Band D calculation of equation (1).

If neither of the S203 and S205 tests is met, then the derivative value of S201 is output to S207. If the derivative value is in the median (middle) of this stored range of values, then in S208 a digital
10 numerical value of Yield D= Median range is output to S211 for use in the Band D calculation of equation (1). If the value of the motor speed derivative does not satisfy any of the S203, S205 and S207 conditions, then in S210 a numerical digital value of Yield D =
15 Low is passed to S211 for the Band D calculation. The calculation of the Band D value in S211 is concluded using equation (1) and this digital numerical value is available in S213 to be used in a manner to be described with respect to figure 4.

20 Referring now to figure 3, S300 shows the determination of a Band P proportional error value that is accomplished in a manner similar to that of calculating the Band D value. In S301, an Error P is calculated as:

25
$$\text{Error P} = \text{Current Speed} - \text{Preset Speed} \quad (2)$$

where the current speed value is provided by the tachometer 5 and the preset speed is a value programmed in the controller 7. The controller 7 is also programmed with a table of values of a Yield P
30 value that corresponds to the calculated values of Error P.

Determination of Yield P and the subsequent calculation of the Band P result are carried out only if the procedure of S201, S203, S205 and S207-S208
35 (figure 2) has determined that the value of Yield D is

in the median value range. If this condition exists, as determined in S303, the Yield D = median value is received from S208 of figure 2. This means that the motor is operating at a speed at which a correction
5 can be made to achieve the preset speed. If this motor speed prevails, then the Error P value determined in S301 is processed further. If the Yield D value is outside of the median value range, that is, Yield D is either 0, high or low, then there is no calculation of
10 the Band P value.

The Error P value that passes in S303 is tested in S305 and S307 to determine if it is of either a low or high value. An Error P value at the low end of the range produces in S306 a Yield P = Low signal of a
15 predetermined digital value that is applied to a Band P calculation step S331. If the S303 determination is that the Error P value is not in the low end of the error P range, then Error P is again tested in S307 to determine if it is in the high end of this range. If
20 it is, then S308 supplies the corresponding numerical digital value of Yield P= High to the Band P calculation step S331.

If the Error P value is neither Low (S305) nor High (S307), then Error P is output to S309 which is also
25 supplied with the motor speed derivative signal value of S201. If the value of the motor speed derivative signal is in the low part of its range, then the Error P value is passed to S311 to determine both if the value of Error P is positive and the motor speed
30 derivative (of S201) is negative. In S309, the low numerical value of the motor speed derivative signal is set directly from the stored table referred to in Figure 2. If in S311 the result is negative, then the Error P value is further tested in S313 to determine
35 both if Error P (S301) is negative and the motor speed

derivative (S201) is positive. A positive result of either of S311 or S313 causes a Yield P= Low numerical value to be outputted in S315 to the Band P calculation step S331. If both S311 and S313 are
5 negative, then the Error P value is passed to S331 for the Band P calculation.

The Band P value is calculated as:

$$\text{Band P} = \text{Error P} \times \text{Yield P} \quad (3)$$

where,

10 Error P is produced in S301, S303, and S305, and Yield P is produced in one of S306, S308 or S315.

The calculation of the Band P value is completed in accordance with equation (3) and is available at S333. Figure 4 shows the production of the signal used to
15 control the triggering of the triac 4 using the Band D and Band P values of figures 2 and 3 (S213 and S331). After the start in S401, the controller 7 in S403 activates its memory and calculating sections to set the current motor speed and stored value of prior
20 speed, both measured by the tachometer 5, to zero. The electrical triggering angle of the triac 4 also is set to zero. In S405, the current motor speed value is supplied by the tachometer and this value is also supplied to S427 where calculation of the triggering
25 angle control for the triac 4 takes place.

The determination and calculation of the Band D and Band P values described with respect to figure 2 (S200) and figure 3 (S300) takes place. These values are added in S407 to produce:

$$\text{Total Error} = \text{Band D} + \text{Band P} \quad (4)$$

30 The Total Error is roughly proportional to the difference that the triac 4 triggering angle deviates from the preset value needed to produce the preset motor speed. Both the Band D and Band P values are
35 related to the difference in actual motor speed

relative to the preset speed.

In S409, the Total Error value of S407 is converted to the electrical angle needed to be used to correct the triac 4 triggering angle to achieve the preset motor speed. If appropriate, the electrical degree triggering angle error value from S409 is limited in S411 to less than a preset value to prevent production of a signal that would result in excessive changes in the triac 4 triggering angle. As noted above, the controller 7 is programmed with a reference numerical digital value that corresponds to the correct triggering angle for the triac 4 to achieve the preset motor speed.

S413 performs the calculation of:

Trig. Angle = Preset Trig. Angle + Error (S411) (5)
which is the final value to be used to control the triac 4 triggering angle. The digital value of the Preset Triggering Angle is programmed in the controller 7. The Triggering Angle value of S413 is limited in S415 to a value that would not cause loss of control of the triac 4.

In S417, the prior motor speed value in S201 is changed to the current speed value to be available for the next motor speed first derivative calculation in S201, also made available in S301.

The output of S413 is applied to a timing control circuit S427 which applies a digital signal to the D/A converter 8 that produces an analog signal that is applied to the triac 4 that controls its triggering angle and therefore the motor speed to obtain the preset speed.

It should be understood that the acquisition of the various data parameters, such as motor speed derivative and speed difference and calculation of the Band D, Band P and total error values can be effected

as often and as many times as necessary. All of this is within the microprocessor 7 operating program, so long as it is within the operating capability of the microprocessor.

5 Specific features of the invention are shown in one or more of the drawings for convenience only, as each feature may be combined with other features in accordance with the invention. Alternative embodiments will be recognized by those skilled in the art and are
10 included within the scope of the appended claims. Accordingly, the above description should be construed as illustrating and not limiting the scope of the invention and all obvious changes and modifications are to be considered within the scope of the appended
15 claims.